

Scalable Interconnection Using a Superconducting Flux Qubit

Daisuke Saida^{1,2*}, Kazumasa Makise^{1,3} and Mutsuo Hidaka¹

¹National Institute of Advanced Industrial Science and Technology

²Fujitsu Limited

³National Astronomical Observatory of Japan

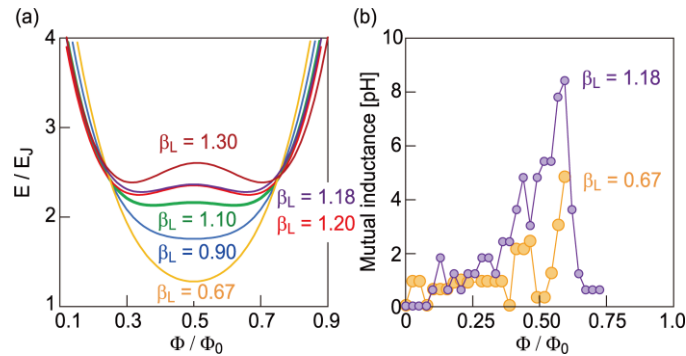
*Corresponding author: saida.daisuke@aist.go.jp, saida.daisuke@fujitsu.com

Supplementary Note 1: Tunable coupler in the superconducting multi-layer quantum circuit

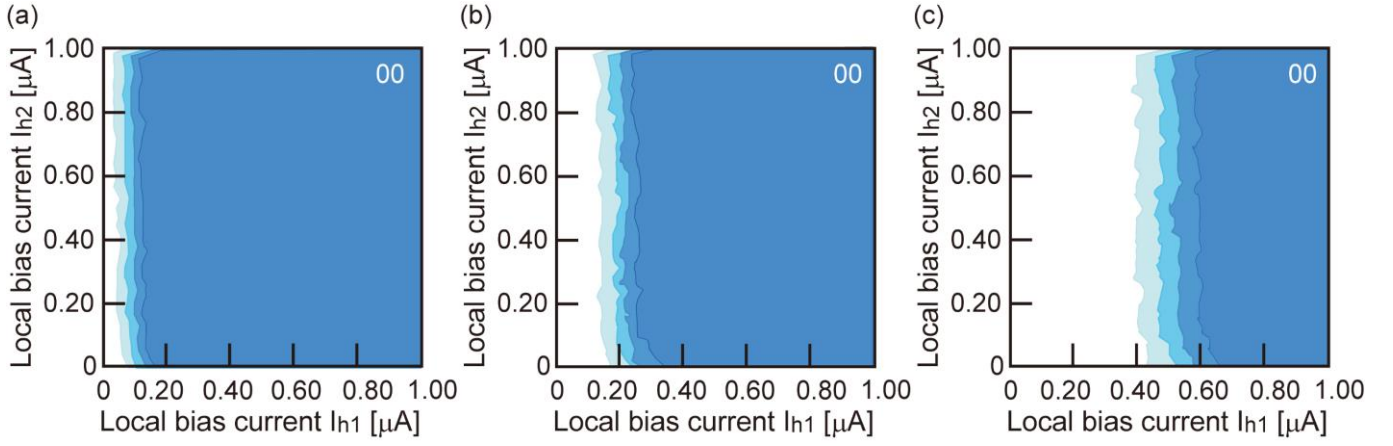
A quantum circuit composed of superconducting multilayer wiring for quantum annealing (QA) uses a tunable coupler for modulating the coupling strength between qubits. The compound Josephson junction (CJJ) rf-superconducting quantum interference device (rf-SQUID) is used for the tunable coupler¹.

The design of the shape of the bottom of the energy potential in the coupler is crucial for accurate tunability. This equates to designing the value of dimensionless factor $\beta_L (= 2\pi LI_c/\Phi_0)$, where L is inductance, I_c is critical current in the JJ, and Φ_0 is the flux quantum) as 0.9–1.2¹ (Supplementary Fig. 1a). A coupler that has an energy potential with an uneven bottom cannot control the coupling magnitude accurately (β_L of 0.67 in Supplementary Fig. 1b). The results shown here are obtained experimentally at 4.2 K. In contrast, the coupler designed with β_L of 1.18 modulates the coupling strength by several picohenry. The energy potential becomes a double-well shape above β_L of 1.2, which should be avoided. Above β_L of 1.2, the coupling conditions cannot be modulated correctly in the conventional method. Conventionally, the coupler works like mutually coupled inductors that can be modulated. In the initial conditions, the flux strength applied to the coupler is fixed before QA of the qubits. The flux strength is set to correspond to an interaction term (J_{ij}) in the Hamiltonian. For QA, mutual inductance (M) in each qubit coupling is initially determined based on the design of J_{ij} . The value of M is tuned by applying the magnetic flux before each QA. However, in the double-well shape of the energy potential, errors occur frequently because of the quasi-stable nearest-neighbor energy state, limiting the range of the flux modulation.

The limitation of the design value of β_L in the tunable coupler is inconvenient because it restricts the interconnection distance, thereby restricting routing in the circuit. Although the value of L in the coupler is adjustable by modulating the width of the main loop, the loop length cannot be sufficiently large. Another problem is the low coupling strength of only several picohenry. Therefore, we use a connection qubit (CQ) for the interconnection between qubits. In the CQ, the maximum value of M , which is designed to be about 10 times larger than that of the tunable coupler, is determined by overlapping areas between the qubit and the CQ. The coupling strength can be tuned time-dependently by applying the magnetic flux.



Supplementary Fig. 1a Energy potential of the tunable coupler based on the CJJ rf-SQUID for various values of β_L . **b** Modulation of the mutual inductance of the tunable coupler measured experimentally at 4.2 K. Purple and orange plots are the results for the tunable couplers with β_L of 1.18 and 0.67, respectively.

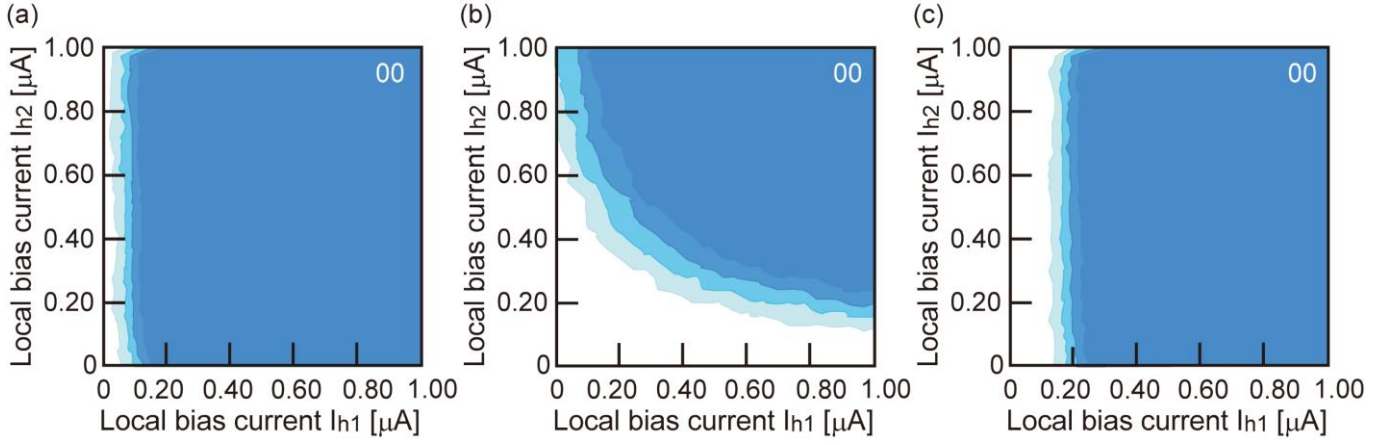


Supplementary Fig. 2 Phase diagram of 00 state with T_a of **a** 100, **b** 15, and **c** 1 μs . The experiment is carried out at 10 mK using CQ-A ($L_{\text{CQ}} = 258$ pH). To clarify the T_a dependence of the 00 state, other regions are not shown. The color shading represents the probability from 0 to 1 that the 00 state occurs.

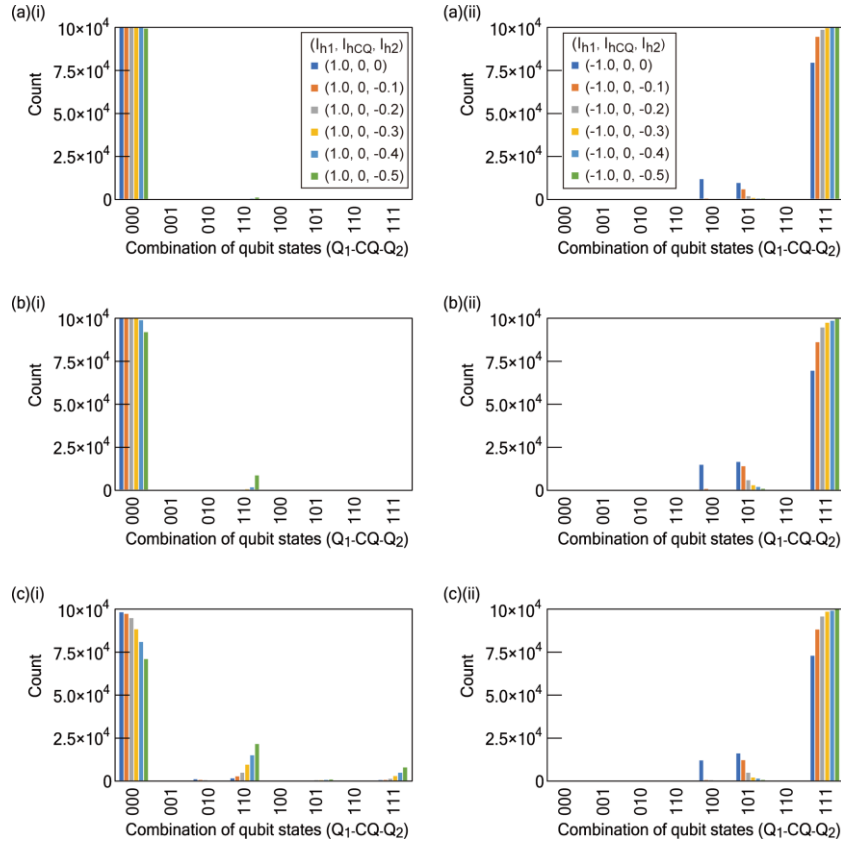
Supplementary Note 2: Characteristics of the phase diagram in two qubits coupled by the CQ

We focus on the region where both qubits (Q1 and Q2) take the 0 state (00 state) in Fig. 3 and discuss its characteristics. Supplementary Figs. 2a, 2b, and 2c show the phase diagram of the 00 state in CQ-A for annealing times (T_a) of 100, 15, and 1 μs , respectively. For simplicity, only the 00 state is shown. The gray zone, which is the transition area between states, increases as T_a decreases. The experiment is mainly carried out at T_a of 100 μs because the in-phase state of two qubits can be observed over a wide range of current conditions. This is also related to the mechanical design of wires through which the voltage that provides the magnetic flux passes. The whole experimental system is designed to propagate voltage signals from several kilohertz to 100 kHz.

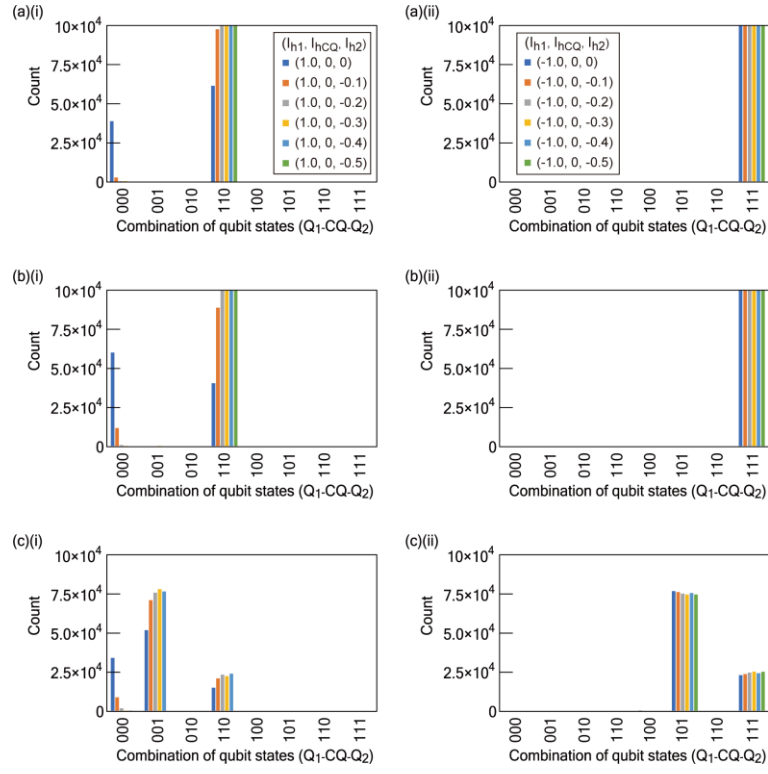
Supplementary Figs. 3a, 3b, and 3c show the phase diagrams of the 00 states in CQ-A, CQ-B, and CQ-C, respectively, at T_a of 100 μs . The local bias current (I_h) condition that gives the 00 state becomes narrower as the value of L in the CQ (L_{CQ}) increases. We focus on different features in the phase diagram of qubits coupled by CQ-B and CQ-C. When the state transition probability of CQ-C is evaluated alone, it shows no monotonic change with I_h . This is mainly due to the shape of the energy potential in the CQ². The metastable state in the energy potential begins to appear above β_L of 8. In CQ-B, the value of β_L is designed to be about 6. Although no metastable state appears in the energy potential, the designed conditions for the qubit are in the boundary where the qubit state does not change monotonically with the flux. Supplementary Figs. 4a, 4b, and 4c show histograms of the combination of qubit states (Q1-CQ-Q2) at T_a of 100, 15, and 1 μs , respectively, for qubits coupled with CQ-A. The 000 and 111 states occur frequently. Supplementary Figs. 5a, 5b, and 5c show histograms of the combination of qubit states (Q1-CQ-Q2) at T_a of 100, 15, and 1 μs , respectively, for qubits coupled with CQ-B. The 110 state occurs frequently in conditions with I_{h1} of 1.0 μA . This means that the 10 region is extended in the phase diagrams of Q1 and Q2. Supplementary Figs. 6a, 6b, and 6c show histograms of the combination of qubit states (Q1-CQ-Q2) at T_a of 100, 15, and 1 μs , respectively, for qubits coupled with CQ-C. Similar to the results for CQ-B, the 110 states occur frequently in conditions with I_{h1} of 1.0 μA . The 111 state also appears with I_{h1} of 1.0 μA , meaning that the 11 region is extended in the phase diagrams of Q1 and Q2. The metastable energy potential of CQ-C may affect these features. In Fig. 4, the generation of the 000 and 111 states cannot be intentionally controlled by using CQ-B and CQ-C. The reason for this is related to the distribution of the identical states in the phase diagram. In QA, the CQ should be designed with β_L from 1.2 to 6 so that the energy potential adopts a clear double-well shape when transverse flux ($\Phi_{\text{trans_CQ}}$) of Φ_0 is applied.



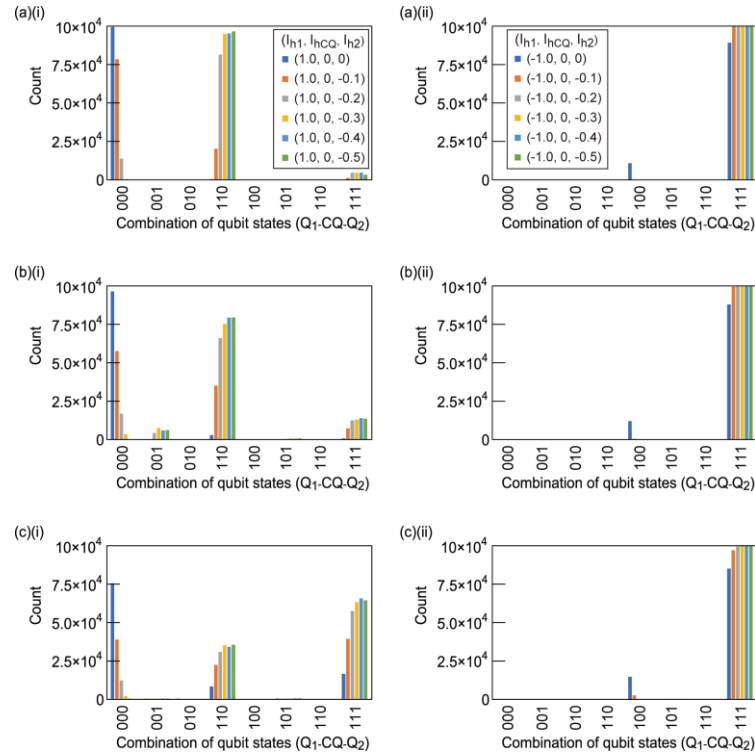
Supplementary Fig. 3 Phase diagram of 00 state in qubits coupled experimentally by **a** CQ-A ($L_{CQ} = 258$ pH), **b** CQ-B ($L_{CQ} = 408$ pH), and **c** CQ-C ($L_{CQ} = 538$ pH) with T_a of 100 μ s at 10 mK. To clarify the L_{CQ} dependence of the 00 state, other regions are not shown. The color shading represents the probability from 0 to 1 that the 00 state occurs. The β_L values of CQ-A, CQ-B, and CQ-C are 3.9, 6.2, and 8.2, respectively. Above β_L of 6.0, the energy potential does not change monotonically with the flux².



Supplementary Fig. 4 Histograms of each qubit state in two qubits coupled by CQ-A after QA with T_a of 100 μ s under current conditions of **a(i)** $I_{h1} = 1.0$ μ A and **a(ii)** $I_{h1} = -1.0$ μ A, with T_a of 15 μ s under current conditions of **b(i)** $I_{h1} = 1.0$ μ A and **b(ii)** $I_{h1} = -1.0$ μ A, and with T_a of 1 μ s under current conditions of **c(i)** $I_{h1} = 1.0$ μ A and **c(ii)** $I_{h1} = -1.0$ μ A. Experiments are carried out at 10 mK. In each experiment, the local bias current of CQ (I_{hCQ}) is kept at 0 μ A. The local bias current of Q2 (I_{h2}) is modulated from -0.5 to 0 μ A.



Supplementary Fig. 5 Histograms of each qubit state in two qubits coupled by CQ-B after QA with T_a of 100 μ s under current conditions of **a(i)** $I_{h1} = 1.0$ μ A and **a(ii)** $I_{h1} = -1.0$ μ A, with T_a of 15 μ s under current conditions of **b(i)** $I_{h1} = 1.0$ μ A and **b(ii)** $I_{h1} = -1.0$ μ A, and with T_a of 1 μ s under current conditions of **c(i)** $I_{h1} = 1.0$ μ A and **c(ii)** $I_{h1} = -1.0$ μ A. Experiments are carried out at 10 mK. In each experiment, I_{hCQ} is kept at 0 μ A. I_{h2} is modulated from -0.5 to 0 μ A.



Supplementary Fig. 6 Histograms of each qubit state in two qubits coupled by CQ-C after QA with T_a of 100 μ s under current conditions of **a(i)** $I_{h1} = 1.0$ μ A and **a(ii)** $I_{h1} = -1.0$ μ A, with T_a of 15 μ s under current conditions of **b(i)** $I_{h1} = 1.0$ μ A and **b(ii)** $I_{h1} = -1.0$ μ A, and with T_a of 1 μ s under current conditions of **c(i)** $I_{h1} = 1.0$ μ A and **c(ii)** $I_{h1} = -1.0$ μ A. Experiments are carried out at 10 mK. In each experiment, I_{hCQ} is kept at 0 μ A. I_{h2} is modulated from -0.5 to 0 μ A.

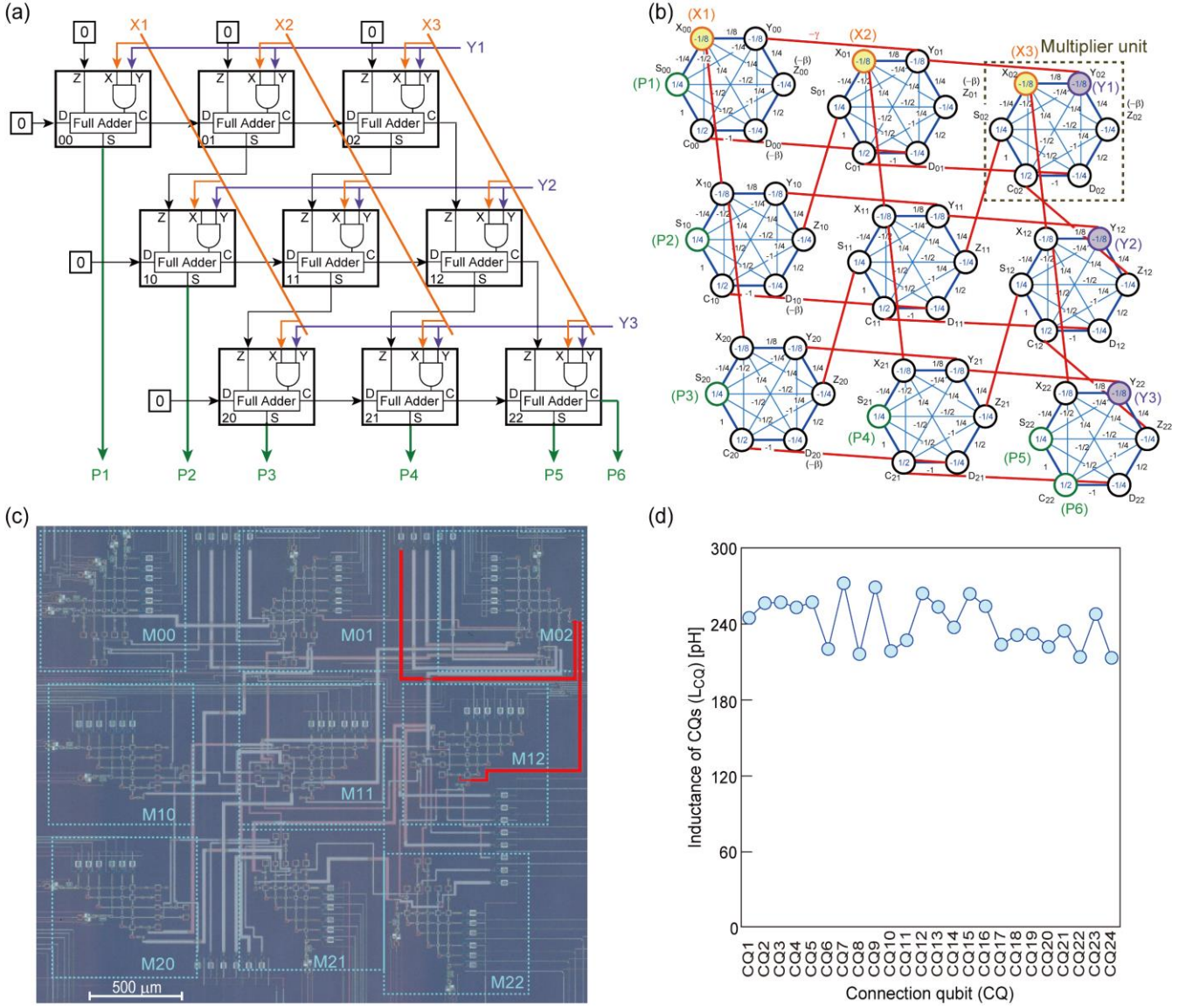
Supplementary Note 3: 6-bit factorization circuit

In QA, we design a problem Hamiltonian with the ground state corresponding to the solution. The desired solution can be obtained by giving a local bias current for initial conditions, and invertible logic³ can also be used. We can perform the prime factorization in QA based on invertible logic by using a classical multiplier Hamiltonian. Supplementary Fig. 7a shows a schematic of the classical 6-bit multiplier circuit. In the multiplication, inputs correspond to the combination of $X = (X_3, X_2, X_1)$ and $Y = (Y_3, Y_2, Y_1)$. Output is $P = (P_6, P_5, P_4, P_3, P_2, P_1)$. Using the ground-state spin logic⁴, the Hamiltonian of this circuit can be expressed as shown in Supplementary Fig. 7b. This Hamiltonian consists of the multiplier unit (MU) and qubit interconnections corresponding to carry propagation path shown in Supplementary Fig. 7a. The interconnection is realized with the CQ. The distances required for each interconnection are different, unlike the couplings between unit lattices in the chimera graph architecture⁵.

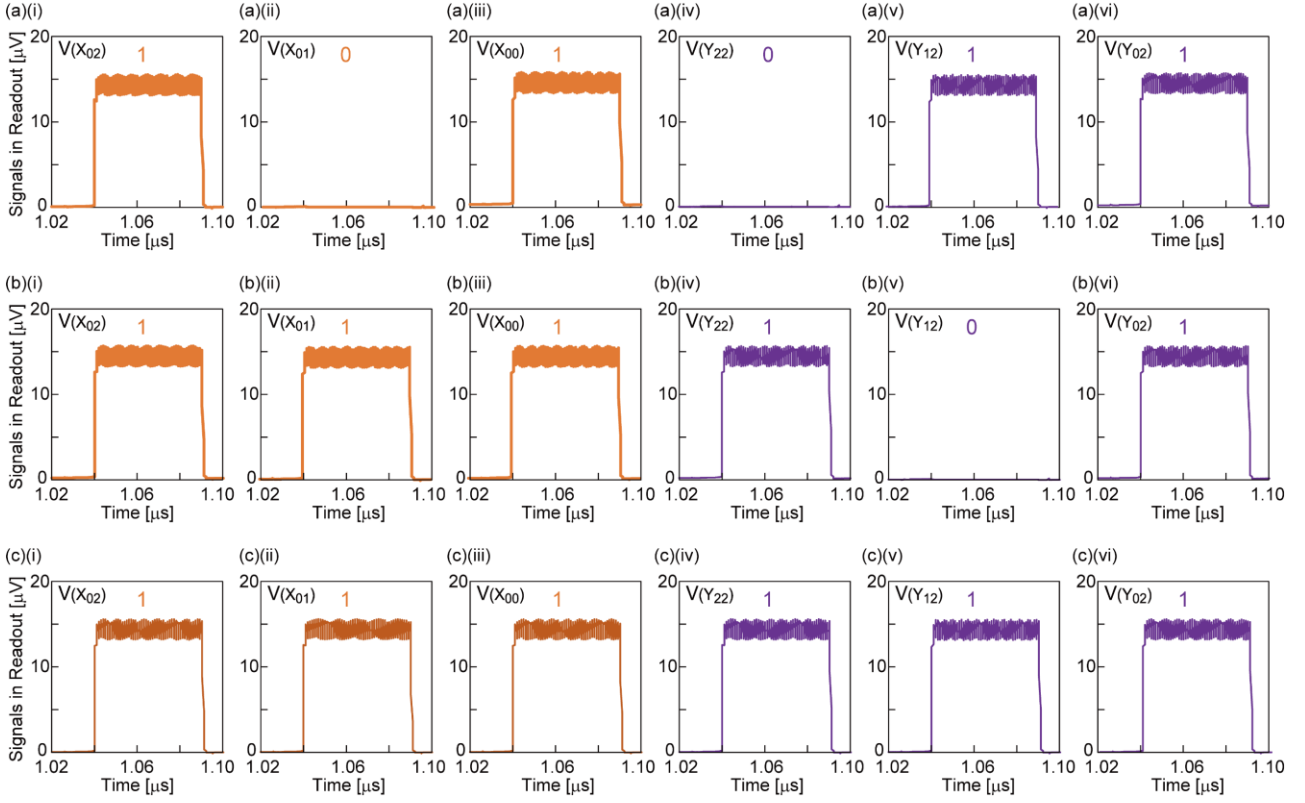
Supplementary Fig. 7c is an optical micrograph of a 6-bit factorization circuit composed of 9 MUs and 24 CQs. The CQ connecting the longest distance is shown in red. The main loop of the CQ is configured so that the outward and return paths are opposite each other. The length of the coupling between the qubits exceeds 7 mm. As discussed in Supplementary Note 2, the design of L_{CQ} is important to prevent errors in the interconnection. The values of L_{CQ} are designed as an average of 241 pH with a standard deviation of $\pm 7\%$ (Supplementary Fig. 7d). Because each CQ has a different main loop length, we adjust both the loop width and use the metal layer to construct the loop. The 6-bit factorization circuit consists of 78 qubits, and 288 qubits are required if we directly implement the Hamiltonian shown in Supplementary Fig. 7b in the chimera graph architecture. Our method, in which the design of the superconducting quantum circuit directly corresponds to the Hamiltonian, can help to reduce the number of qubits needed, moving closer to practical use of QA.

Supplementary Figs. 8a, 8b, and 8c show the prime factorization of integers 15, 35, and 49, respectively, in simulations. In this analysis, the annealing is performed with the modulation of the transverse magnetic flux from 0 to Φ_0 during 0.02–1.02 μs . If the signal in the readout has a finite voltage ($>10 \mu\text{V}$) around 1.06 μs , the qubit takes the 1 state. $V(X_{02})$, $V(X_{01})$, and $V(X_{00})$ correspond to X components X_3 , X_2 , and X_1 , respectively. $V(Y_{22})$, $V(Y_{12})$, and $V(Y_{02})$ correspond to Y components Y_3 , Y_2 , and Y_1 , respectively. The prime factors are obtained for each integer. For example, in the prime factorization of integer 15, combinations of $X = (101)_{(2)}$ and $Y = (011)_{(2)}$ or vice versa are obtained. This proves that the CQ combines the unit cell properly to express the entire Hamiltonian.

The CQ enables the quantum circuit to be extended in two dimensions by combining the unit lattice. Because the connection distance varies for each CQ, the individual design of L_{CQ} is necessary. In this study, L_{CQ} was designed with InductEX⁶ for the proof-of-concept demonstration; however, design automation is necessary to increase scalability. In semiconductor circuits, automated design is used for routing the wiring. A similar approach would be suitable for the layout of quantum circuits. In future work, automated design could be treated as a constrained problem with wire width as a parameter, or machine learning could be used.



Supplementary Fig. 7 **a** Schematic of the classical 6-bit multiplier based on MUs (MU_{ij} , $i, j = 0-2$). **b** Description of the Hamiltonian in the 6-bit factorization circuit. Production can be described as $P = (P_6 P_5 P_4 P_3 P_2 P_1)_{(2)} = (C_{22} S_{22} S_{21} S_{20} S_{10} S_{00})_{(2)}$. Inputs are represented as $X = (X_3 X_2 X_1)_{(2)} = (X_{02} X_{01} X_{00})_{(2)}$ and $Y = (Y_3 Y_2 Y_1)_{(2)} = (Y_{22} Y_{12} Y_{02})_{(2)}$, respectively. **c** Optical photograph of a superconducting quantum circuit embedding the 6-bit factorization Hamiltonian. The circuit is composed of 9 MUs and 24 CQs, and consequently uses 78 superconducting flux qubits. **d** Designed L_{CQ} values of the 24 CQs used in the 6-bit factorization circuit within the range of 241 ± 7 pH.



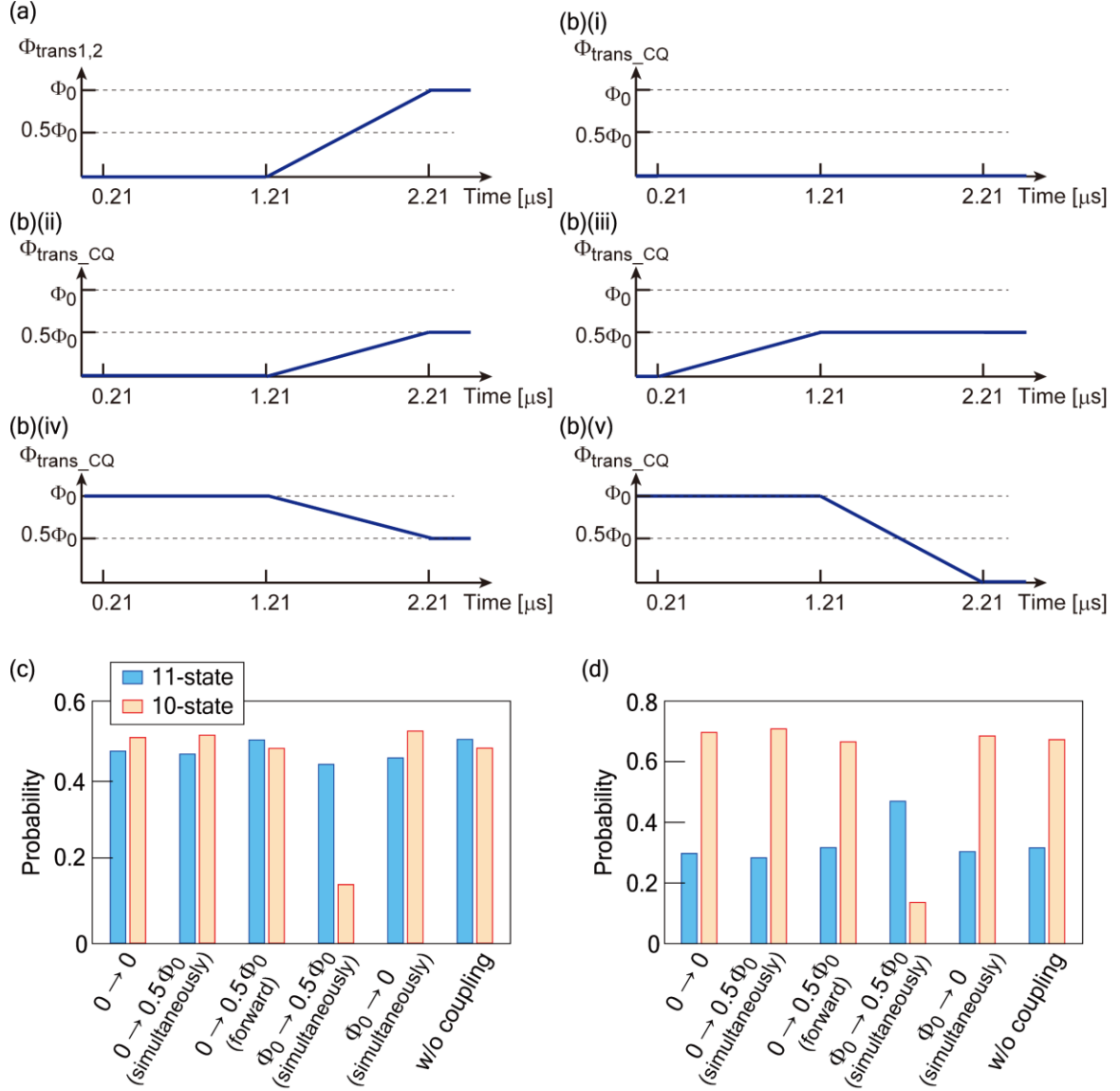
Supplementary Fig. 8 Output signals in readouts for prime factorization of integers P of **a** 15, **b** 35, and **c** 49 in JSIM analysis. Prime factor X is expressed in **(i)** X_3 , **(ii)** X_2 , and **(iii)** X_1 . Prime factor Y is expressed in **(iv)** Y_3 , **(v)** Y_2 , and **(vi)** Y_1 .

Supplementary Note 4: Turning off the qubit coupling via the CQ in QA

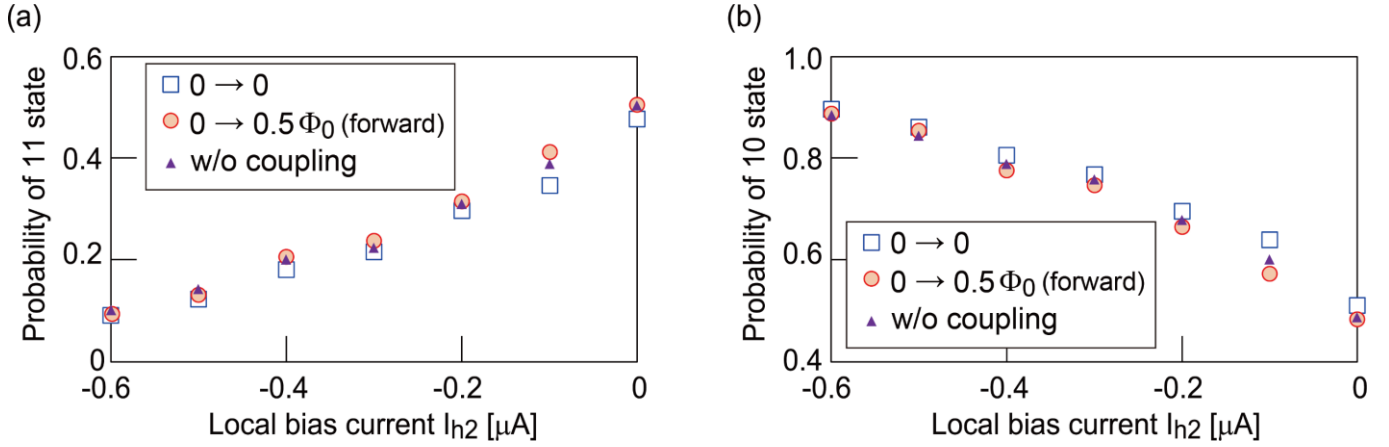
The CQ inductively couples qubits. In the circuit shown in Supplementary Fig. 7, all unit cells (MU_{ij}) are used to implement the problem Hamiltonian for analysis of the prime factorization. Even if a unit lattice is unused, unintended crosstalk can occur during the calculation because other unit lattices are inductively coupled via the CQ. Therefore, it is important to know how to turn off the coupling.

We consider a model of two qubits (Q1 and Q2) coupled by the CQ. The transverse magnetic fluxes (Φ_{trans_i} , $i = 1, 2$) are applied to Q1 and Q2 with modulation from 0 to Φ_0 over 1.21–2.21 μs (Supplementary Fig. 9a). Five ways of applying the transverse magnetic flux ($\Phi_{\text{trans_CQ}}$) to the CQ are investigated. In the first, $\Phi_{\text{trans_CQ}}$ is 0 during the annealing in Q1 and Q2 (Supplementary Fig. 9b(i)). This is equivalent to no flux being applied to the CQ connecting the unit lattice that is not used. In the second and third ways, $\Phi_{\text{trans_CQ}}$ is modulated from 0 to $0.5\Phi_0$. The two ways differ in whether the $\Phi_{\text{trans_CQ}}$ is applied at the same time (Supplementary Fig. 9b(ii)) or earlier than Φ_{trans_i} to Q1 and Q2 (Supplementary Fig. 9b(iii)). In the fourth way, $\Phi_{\text{trans_CQ}}$ is modulated from Φ_0 to $0.5\Phi_0$ and is applied to the CQ at the same time as Φ_{trans_i} on the time axis (Supplementary Fig. 9b(iv)). In the fifth way, $\Phi_{\text{trans_CQ}}$ is modulated from Φ_0 to 0 and is applied to the CQ at the same time as Φ_{trans_i} on the time axis (Supplementary Fig. 9b(v)). To favor Q1 adopting the 1 state, the flux is applied through a local bias current (I_{h1}) of 1.2 μA . Calculations are performed under two current conditions for Q2. We use two I_{h2} values for which Q2 takes the 0 state with a probability of 50% (Supplementary Fig. 9c) and 70% (Supplementary Fig. 9d) when Q2 is isolated. The condition “w/o coupling” is based on the calculation results when Q1 and Q2 are isolated. Strictly off-state coupling between qubits means that the probability is the same as the condition “w/o coupling”. The probability is different between conditions “0→0” (corresponding to constant $\Phi_{\text{trans_CQ}}$ of 0) and “w/o coupling”, indicating that crosstalk occurs via the CQ.

The coupling is strictly in the off-state when $\Phi_{\text{trans_CQ}}$ is varied from 0 to $0.5\Phi_0$ before applying Φ_{trans_i} to Q1 and Q2 (“0→0.5 Φ_0 (forward)” in Supplementary Fig. 10). The energy potential of the CQ becomes a single well, existing at the lowest energy when the $\Phi_{\text{trans_CQ}}$ of $0.5\Phi_0$ is applied. When $\Phi_{\text{trans_CQ}}$ is modulated from 0 to $0.5\Phi_0$, the coupling is more strictly turned off if $\Phi_{\text{trans_CQ}}$ is applied before Φ_{trans_i} compared with applying $\Phi_{\text{trans_CQ}}$ simultaneously with Φ_{trans_i} . This suggests that the magnitude of the energy between the three components affects the occurrence of tunneling. In other words, a large energy separation between qubits and the CQ prevents quantum tunneling and turns off the coupling during annealing.



Supplementary Fig. 9 **a** Time-dependent transverse magnetic flux $\Phi_{\text{trans}1,2}$ applied to qubits Q1 and Q2. **b** Time-dependent transverse magnetic flux $\Phi_{\text{trans_CQ}}$ applied to the CQ with $\Phi_{\text{trans_CQ}}$ **(i)** constant at 0, **(ii)** modulated from 0 to $0.5\Phi_0$ during 1.21–2.21 μs , **(iii)** modulated from 0 to $0.5\Phi_0$ during 0.21–1.21 μs , **(iv)** modulated from Φ_0 to $0.5\Phi_0$ during 1.21–2.21 μs , and **(v)** modulated from Φ_0 to 0 during 1.21–2.21 μs . Probability of generating each qubit state in the two qubits coupled by CQ-A after annealing under current conditions of **c** $I_{h2} = 0 \mu\text{A}$ and **d** $I_{h2} = -0.2 \mu\text{A}$ in the JSIM analysis.



Supplementary Fig. 10 **a** Probability of the 11 state with respect to modulation of I_{h2} . **b** Probability of the 10 state with respect to modulation of I_{h2} . $\Phi_{\text{trans_CQ}}$ is kept at 0 or annealing is performed with $0.5\Phi_0$ before the annealing for Q1 and Q2. The w/o coupling conditions indicate that qubits Q1 and Q2 are independently annealed without the interconnection.

Supplementary Note 5: Reliable connection with high yield in flip-chip bonding technology

We establish a flip-chip bonding process using In-Sn reflow soldering, in which 10- μm -high solder bumps are formed uniformly over the entire wafer surface with high reproducibility. Alignment compression using a semi-auto bonder creates reliable connections with high yield. We use In-Sn alloy as the solder because it has a relatively high critical temperature (T_c) of 5 K, which allows us to confirm the superconducting connection between bonded chips by using a helium dewar vessel or conventional refrigerators.

Supplementary Fig. 11 shows the procedure for forming In-Sn solder bumps on the qubit or readout circuits. First, an adhesive region of Au/Ti is formed on the circuit through the lift-off process. Solder bumps are fabricated by dipping the substrate in molten In-Sn alloy in a solder bath. Transition temperature as a function of composition has been reported in the In-Sn binary system⁷. Amorphous alloy bumps (quenching from liquid state) show higher transition temperatures than crystalline alloy bumps. Equilibrium T_c values varied from 3.4 K (near pure In) to 6.6 K (beta phase, 33.5 at.% Sn).

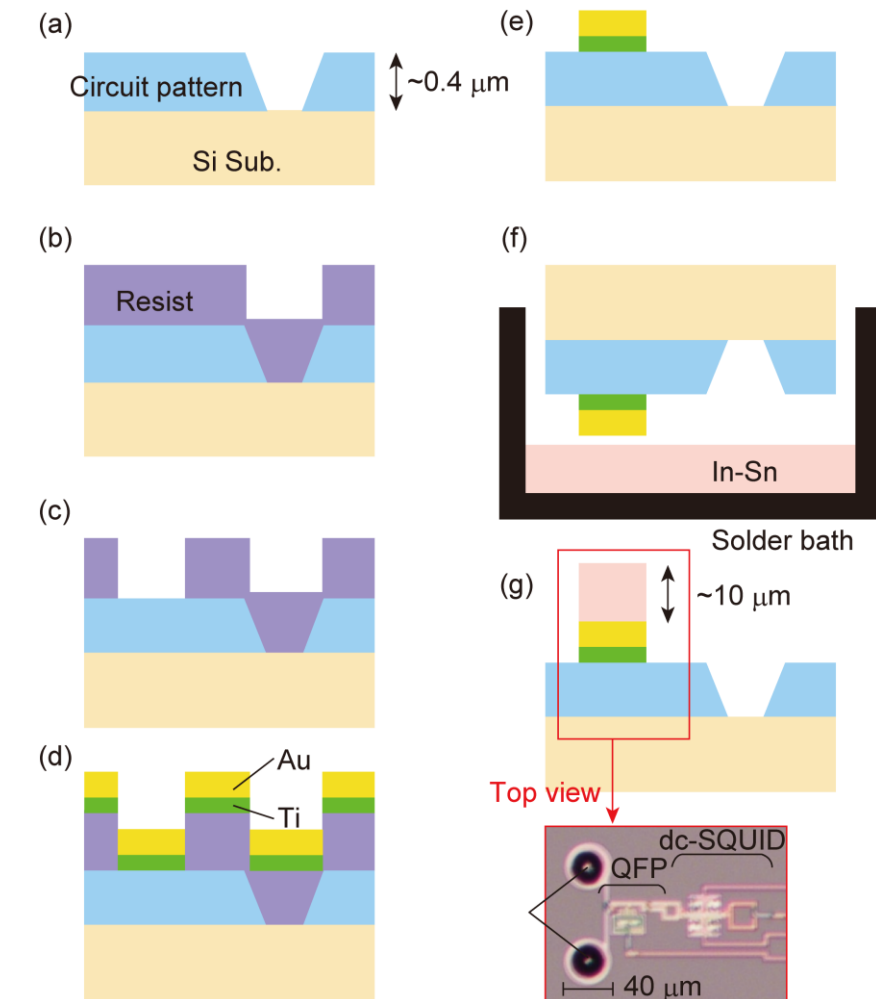
The top and bottom chips are 3.5×3.5 and 7.1×7.1 mm squares, respectively. The adhesive Au/Ti patterned regions have designed diameters of 40 and 100 μm in the fine pattern area in the qubit (readout) and in the surrounding chip area, respectively. This increases stability in the bump connections and it ensures thermal conductivity between the two chips. Thermal conductivity is crucial because the upper chip is cooled through the bumps when the sample is placed in the refrigerator with the bottom chip in contact with a cold plate (heat sink). Supplementary Figs. 12a and 12b show optical micrographs of the top chip (qubit) and bottom chip (readout). The black circles correspond to bumps. Bumps are selectively formed on the adhesive region even though the entire chip is immersed in solder solution.

After the bump formation shown in Supplementary Fig. 11g, the upper chip is flipped so that its surface with the circuit pattern is downward. The bump positions of top and bottom chips are aligned and the chips are compressed using a flip-chip bonder (BFC-1000, Sony EMCS Corporation). Bonding is performed below 150 $^{\circ}\text{C}$ with a typical bonding force of 10–20 N in the bump area (~ 10 g/bump), which results in a compression of roughly 10 μm of the total height of the two solder bump depositions. Infrared microscopic inspection of the top and bottom chips indicates that alignment deviation in the plane is less than ± 5 μm . As a proof-of-concept demonstration, the bump diameter is 40 μm in the fine pattern in this study, although the bump diameter can be reduced to about 10 μm . The rework process is easy because bumps can be made reproducibly by dipping the chip back into the solder solution. The processes after Supplementary Fig. 11(e) can be performed by sputtering

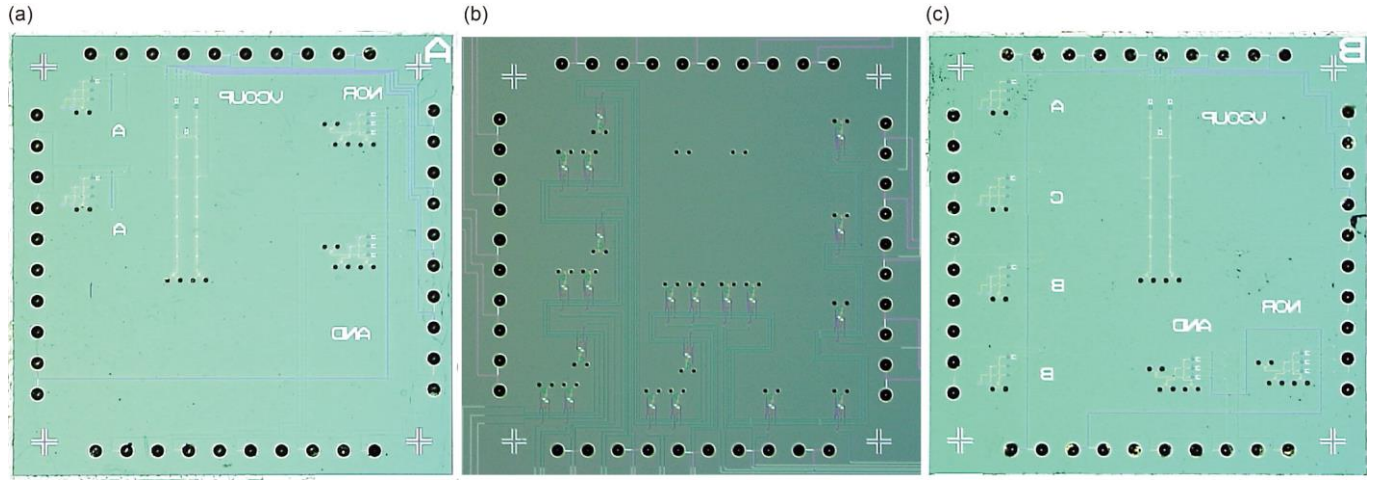
and lift-off method. In this case, the controllability of bump shape will be improved. In addition, due to suppression of heat process, changes in the characteristics of the JJ are avoided. This will be addressed as future work.

A Nb wiring circuit connected in series with 90 solder bumps (daisy-chain structure) was fabricated by flip-chip bonding to verify the superconducting connections. The diameter of the bumps was 100 μm . The fabrication process was similar for the daisy-chain Nb wire sample and the sample shown in Fig. 6b. The daisy-chain Nb wire sample was mounted on a refrigerator and cooled through the bottom chip. Supplementary Fig. 13 shows the resistance–temperature characteristics. The metallic behavior above 10 K is for Nb wiring. The Nb wire undergoes a superconducting transition around 8.6 K. The plateau in the resistance from 6 to 8 K is due to the bumps. Although the effect of contact resistance is included, dividing this value by the number of bumps gives the resistance per unit bump. The resistance is suppressed around 5 K where the bumps undergo the superconducting transition. This result indicates that the upper Nb wire is cooled via bumps, resulting in the whole circuit exhibiting superconducting behavior.

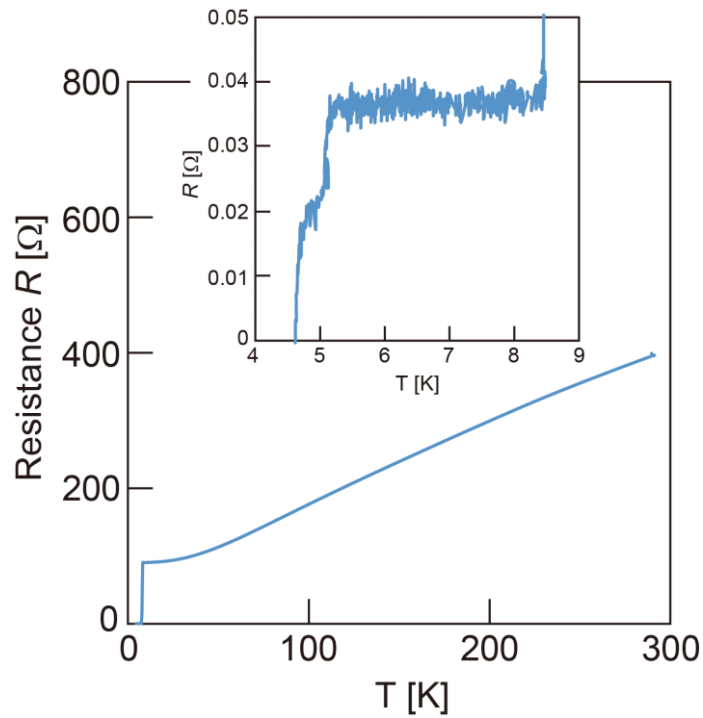
In our method, the bottom chip (readout) can be used in common among different flip-chip bonding configurations by changing the circuits on the top chip (qubit), enabling circuits with different functions to be realized. For example, the top chips shown in Supplementary Fig. 12a and 12c can each be combined with the same bottom chip (Supplementary Fig. 12b).



Supplementary Fig. 11 Procedure for forming an In-Sn solder bump. **a** Initial sample structure. For simplicity, the cross section of a circuit consisting of multiple layers is shown for a single pattern. **b** Resist coating, **c** resist patterning by lithography, **d** sputtering of Ti/Au layers, **e** lift-off, **f** reflow soldering, and **g** formation of In-Sn bumps.



Supplementary Fig. 12 Top-view optical photographs of **a** a 3.5×3.5 mm square qubit chip and **b** a 7.1×7.1 mm square readout chip. In-Sn solder bumps are formed in the areas visible as black circles. **c** Top-view optical photograph of the 3.5×3.5 mm square qubit chip implementing different function compared with that in **a**. The black stuff between the fourth and fifth bumps on the far right is dust.

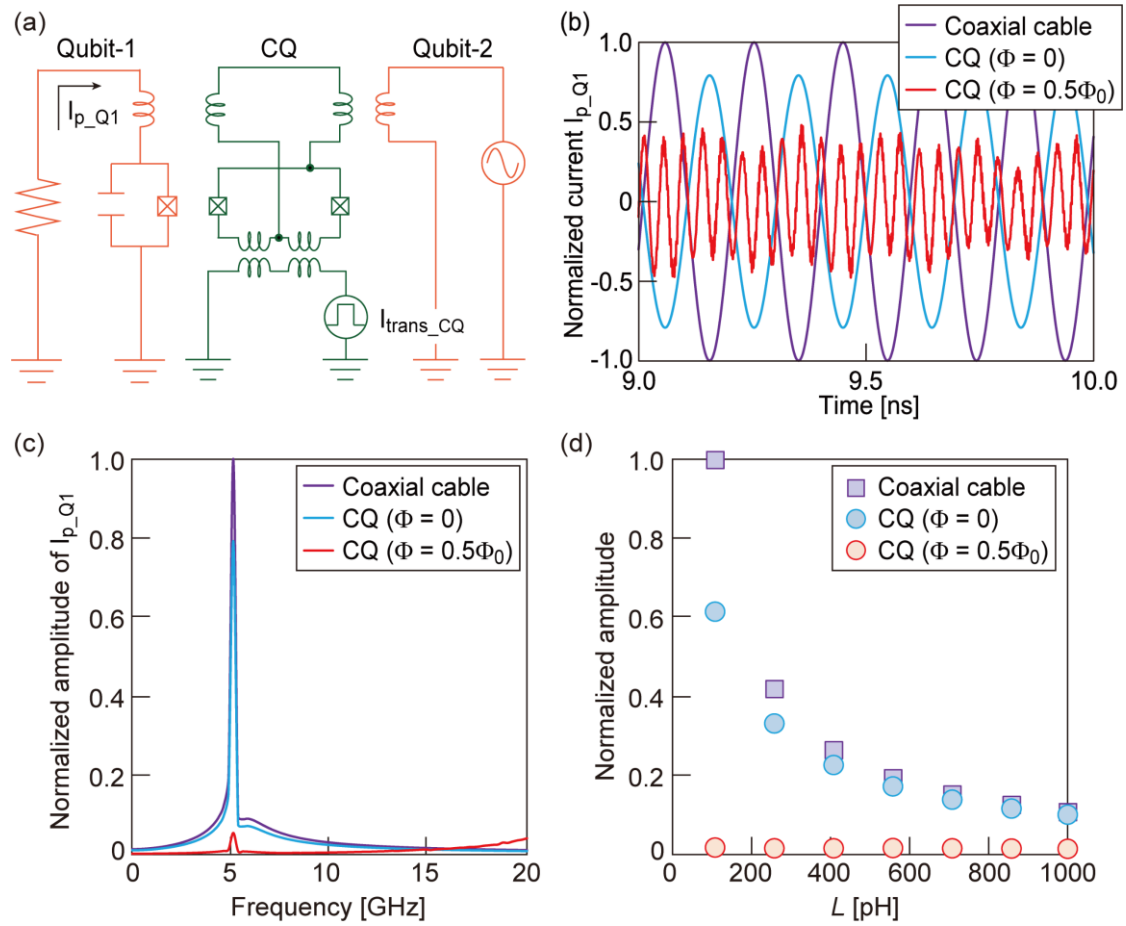


Supplementary Fig. 13 Temperature dependence of the resistance of a circuit composed of Nb wires connected by In-Sn bumps. The Nb wiring circuit was connected in series with 90 solder bumps.

Supplementary Note 6: Interconnect of transmon qubits by the CQ

We assume that two transmon qubits^{10,11} couple through the CQ, as shown in Fig. 7a. Specifically, we consider the transmon qubits formed around the edge of different chips (Qubit Chip-1 and Qubit chip-2). Qubit chip-1, Qubit chip-2, and CQ chip are aligned so that the pattern of each transmon qubit and the main loop in the CQ overlap partially with each other. The ground planes of Qubit chip-1, Qubit chip-2, and CQ chip are connected through bumps by flip-chip bonding. Figure 7b shows the equivalent circuit. Sunada et al.⁹ analyzed external coupling between a transmon qubit and the environment. In that study, the JJ was replaced by an AC current source for analyzing the real part of the admittance in the finite element model. We focus on verifying the possible interaction between transmon qubits coupled by the CQ. In the simulation, a Josephson integrated circuit simulator (JSIM)⁸ is used. Through evaluating the internal current (I_{p_Q1}) in Q1, we examine whether a change of status in Q2 affects Q1 via the CQ. The transmon qubit is assumed to be in the ground state with a zero-point oscillation. We consider the connection status of the transmon qubits under conditions that do not disturb each state. The JJ in the transmon qubit has a resonant frequency of 5.1 GHz and I_c of 30 nA^{10,11}. Here, I_{p_Q1} in the zero-point oscillation is estimated as about 12 nA (about 15 nA in Xmon described in the main manuscript). We consider the interaction between Q1 and Q2 when I_{p_Q2} with a resonant frequency of 5.1 GHz, magnitude of 10 nA, and a time average of zero is generated in Q2. To simulate this, Q2 is replaced by the AC current source. Although Q1 is open-ended, the circuit is shunted by a resistance equivalently. The transmon qubit and the CQ have both capacitive and inductive couplings because parts of the circuit face each other. As a proof-of-concept verification of the interaction, the inductive coupling is considered in the simulation. In Supplementary Fig. 14a, the transmon qubit and CQ are shown in orange and green, respectively, and I_c in the CQ is 2.5 μ A. The interconnection is analyzed while L_{CQ} in the CQ is modulated. For reference, a model of transmon qubits coupled by a coaxial cable with the same L value is also analyzed.

Supplementary Fig. 14b shows the time waveform of I_{p_Q1} for interconnections with the coaxial cable and with the CQ with L of 258 pH. I_{p_Q1} is normalized by the amplitude for the coaxial cable. Currents with the same frequency flow in Q1 for the interconnections with the coaxial cable and with the CQ, where flux is not applied ($\Phi = 0$). In contrast, when flux corresponding to $0.5\Phi_0$ is applied to the CQ, the frequency of I_{p_Q1} changes. The spectra are shown in Supplementary Fig. 14c. The change in resonance frequency of the transmon qubit (5.1 GHz) shows that the coupling status between Q1 and Q2 is switched by the flux through I_{trans_CQ} . The signal intensity without flux is similar to that of the coaxial cable with the same value of L , indicating that the CQ functions as a transmission line with switching characteristics. Supplementary Fig. 14d shows the switching characteristics with respect to L of the interconnection. When flux corresponding to $0.5\Phi_0$ is applied, the interaction is almost completely suppressed at all L values. Similar to the coaxial cable, the magnitude of the interaction depends on L without flux. In our simulations, there was an interaction between two qubits connected by a CQ. It remains an open question whether this interaction would excite the qubit state in a real system, and thus experimental confirmation is needed. In future work, we will verify the coupling of transmon qubits by using the CQ for long interconnections experimentally by actually fabricating a sample. We anticipate that this will pave the way for new quantum circuits using combinations of different types of qubits.



Supplementary Fig. 14 a Circuit model of the transmon qubits coupled by the CQ. **b** Time waveforms of persistent current (I_{p_Q1}) in Qubit-1. Purple line shows I_{p_Q1} for the analysis in transmon qubits coupled by a coaxial cable with L of 258 pH. Blue and red lines show I_{p_Q1} for the analysis in transmon qubits coupled by the CQ without and with applying flux through current I_{trans_CQ} , respectively. The CQ has L of 258 pH and a JJ with I_c of 2.5 μ A. **c** Frequency spectra of I_{p_Q1} . **d** Switching characteristics with respect to the L value of the interconnection.

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